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<div>(84) Designated Contracting States: AT BE CH CY DE DK ES FI FR GB GR IE IT LI LU MC NL PT SE Designated Extension States: AL LT LV MK RO SI</div> <div>(71) Applicant: Koninklijke Philips Electronics N.V. 5621 BA Eindhoven (NL)</div> <div>(72) Inventors: • Smidt, Pieter Jan Mark Prof. Holstlaan 6, 5656 AA Eindhoven (NL)</div>	<div>• Tolle, Tobias Prof. Holstlaan 6, 5656 AA Eindhoven (NL)</div> <div>(74) Representative: Steenbeek, Leonardus Johannes INTERNATIONAAL OCTROOIBUREAU B.V., Prof. Holstlaan 6 5656 AA Eindhoven (NL)</div>
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(54)

Power supply circuit

(57) In a power supply circuit, comprising a rectifier unit (D) for rectifying an AC mains voltage (V_m), a thyristor (T) coupled across first and second outputs of the rectifier unit (D), and a thyristor control circuit (T1, T1', Rb) for controlling the thyristor (T) in such a manner, that the thyristor (T) is switched on when a voltage over the

thyristor (T) is substantially zero, the thyristor control circuit (T1, T1', Rb) includes a transistor (T1, T1') having a main current path between a control terminal of the thyristor (T) and the first output of the rectifier unit, and a resistor (Rb) between a control terminal of the transistor (T1, T1') and the second output of the rectifier unit (D).

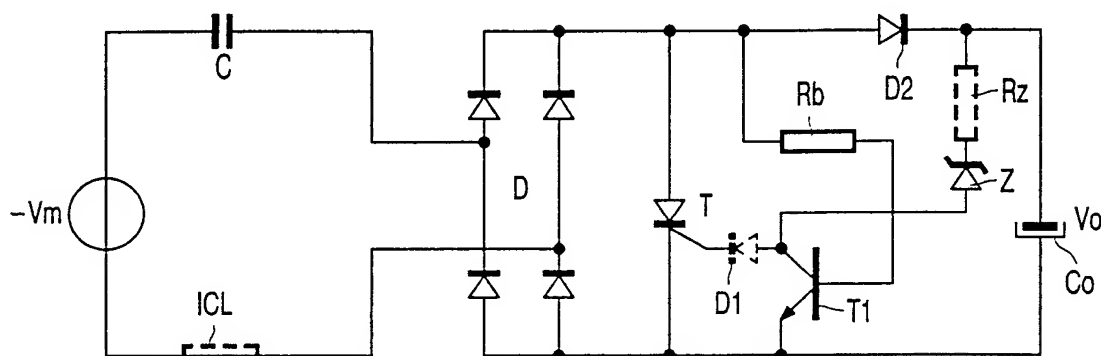


FIG. 1

Description

[0001] The invention relates to a power supply circuit.

[0002] EP-A-0,030,266 discloses a transformer-less power supply that is used to supply a reduced DC operating voltage from an AC supply voltage to an electronic device with a low current requirement. The AC supply voltage is connected via a reactance to a two-way rectifier. The rectifier supplies the DC power via a diode to a storage capacitor that, in turn supplies power to the electronic device. A controlled switch in the form of a thyristor is connected in parallel to the output of the two-way rectifier. The thyristor, which is controlled by a zener diode, is arranged to close when a predetermined voltage at the storage device is exceeded thus short-circuiting the rectifier and isolating the capacitor from the power supply. As soon as the current through the thyristor drops below the self-holding value, the thyristor opens once more and the storage capacitor is once again charged by the output from the two-way rectifier. Switching the thyristor on at a voltage exceeding a predetermined voltage is disadvantageous from an interference and dissipation point of view.

[0003] EP-B-0,324,904 shows a power supply circuit in which the controlled switch is a power transistor. This power transistor is controlled by a control device in such a manner, that at the switching time of the semiconductor switch, the product of current times voltage is always extremely low, so that radio interference is avoided. In switching into the current-conducting state at the end of a voltage half-wave, both current and voltage are very low. At the moment of switching into the off state, the current is high but the voltage is very low so that only slight dissipation occurs. The switching-on time instants for the semiconductor switch may be effected depending on the AC voltage fed to the rectifier bridge circuit, with for example the zero passages determining the switching time. The positive pole of the rectifier bridge circuit is connected to the input of a first threshold stage (a Schmitt trigger), to which a very low reference voltage is applied. The output of the first threshold stage is connected to the setting input of a flip-flop, the output of which is connected to the control input (base) of the semiconductor switch (transistor). A diode is connected between the positive pole of the rectifier bridge circuit and the output capacitor to prevent discharge of the output capacitor via the semiconductor switch. The output voltage at the output capacitor is fed to an input of a second threshold stage (also a Schmitt trigger), the output of which is connected to the resetting input of the flip-flop. A second reference voltage corresponding to a preset setpoint value is applied to the second threshold stage. To prevent current overload of the rectifier assembly, at least a resistor is connected between the mains connection and the semiconductor switch.

[0004] It is, inter alia, an object of the invention to provide an improved power supply circuit, in which switching at a low voltage is achieved by means of a very simple

control circuit. To this end, the invention provides a power supply circuit as defined in the independent claim. Advantageous embodiments are defined in the dependent claims.

[0005] In a preferred embodiment of the power supply circuit, comprising a rectifier unit for rectifying an AC mains voltage, a thyristor coupled across first and second outputs of the rectifier unit, and a thyristor control circuit for controlling the thyristor in such a manner, that the thyristor is switched on when a voltage over the thyristor is substantially zero, the thyristor control circuit includes a transistor having a main current path between a control terminal of the thyristor and the first output of the rectifier unit, and a resistor between a control terminal of the transistor and the second output of the rectifier unit.

[0006] These and other aspects of the invention will be apparent from and elucidated with reference to the embodiments described hereinafter.

[0007] In the drawings:

Fig. 1 shows a first embodiment of a power supply circuit in accordance with the present invention;
Fig. 2 shows a second embodiment of a power supply circuit in accordance with the present invention;
Fig. 3 shows a third embodiment of a power supply circuit in accordance with the present invention; and
Fig. 4 shows a fourth embodiment of a power supply circuit in accordance with the present invention.

[0008] Fig. 1 shows a first embodiment of a power supply circuit in accordance with the present invention. A mains voltage source V_m is connected to a diode bridge rectifier D thru a capacitor C and an optional inrush current limiter ICL. A thyristor T is connected across outputs of the diode bridge rectifier D. A series connection of a diode D2 and an electrolytic capacitor C_o as also connected across outputs of the diode bridge rectifier D. An output voltage V_o is present across the electrolytic capacitor C_o .

[0009] In accordance with the present invention, a simple control circuit is used to control the thyristor T in dependence upon the output voltage V_o . The control circuit includes a bipolar NPN transistor T1 having an emitter connected to a lower terminal of the electrolytic capacitor C_o . A collector of the transistor T1 is connected to a gate of the thyristor T thru an optional diode D1, and to an upper terminal of the electrolytic capacitor C_o thru a zener diode Z and an optional resistor R_z . A base of the transistor T1 is connected an anode of the thyristor T thru a resistor R_b .

[0010] Fig. 2 shows a second embodiment of a power supply circuit in accordance with the present invention. This second embodiment differs from the first embodiment in that instead of the zener diode Z and the resistor R_z , a series connection of two resistors R1 and R2 is present. This series connection is connected across the output capacitor C_o . A connection point of the resistors

R1 and R2 is connected to the collector of the transistor T1.

[0011] Fig. 3 shows a third embodiment of a power supply circuit in accordance with the present invention. This third embodiment is a dual version of the first embodiment. Instead of an NPN transistor T1, a PNP transistor T1' is present and all connections within the control circuit are altered in a corresponding manner.

[0012] Fig. 4 shows a fourth embodiment of a power supply circuit in accordance with the present invention. This fourth embodiment is a dual version of the second embodiment. Instead of an NPN transistor T1, a PNP transistor T1' is present and all connections within the control circuit are altered in a corresponding manner.

[0013] The working is as follows. Normally the 50 Hz capacitor current is charging the output electrolytic capacitor Co via diode D2. If the output voltage Vo reaches its required level, the controlled switch (in this case, a thyristor T) is switched on. From this moment, the capacitor current freewheels through the switch T, not causing any losses. If the output requires power again, the switch T will be opened. A full bridge rectifier D is used, allowing for utilization of both of the mains cycles.

[0014] The control of the switch T can be in such a way that only for full mains cycles the switch T is on or off (with a thyristor this is always the case), thus preventing oscillations. The other reason why a thyristor is preferred, is that it automatically goes out of conduction when its current is zero. This eliminates switch-off losses in the thyristor T and minimizes the EMC problems caused by switching-off.

[0015] By also switching on the thyristor T when its voltage is zero, the switch-on losses are eliminated too as well as the EMC problems that come with it. To this end, in accordance with the present invention, the thyristor T is controlled in such a manner that it is switched on when its voltage is zero. The result is an ultra low-cost capacitive down-converter that has very low losses even at low load such as at standby, and an optimized EMC behavior.

[0016] The transistor T1 keeps the control voltage of the thyristor T low until the voltage over the thyristor T reaches zero; thereafter the thyristor T is allowed to go into the conducting state. The thyristor T is "off" until the transistor T1 is "off". The output voltage Vo over the output capacitor Co is determined by the zener diode Z or by means of a voltage division resulting from the use of the series resistors R1, R2. The optional diode D1 serves to increase the threshold voltage of the thyristor T from 1.0 V to 1.7 V. The optional resistor Rz serves to limit the current thru the zener diode Z.

[0017] It should be noted that the above-mentioned embodiments illustrate rather than limit the invention, and that those skilled in the art will be able to design many alternative embodiments without departing from the scope of the appended claims. The thyristor T may be replaced by any other active switch, such as a (MOS) (field-effect) transistor. The (bipolar) transistor T1 may

be replaced by any other active switch, such as a (MOS) field-effect transistor or a thyristor. In the claims, any reference signs placed between parentheses shall not be construed as limiting the claim. The word "comprising" does not exclude the presence of elements or steps other than those listed in a claim. The word "a" or "an" preceding an element does not exclude the presence of a plurality of such elements. In the device claim enumerating several means, several of these means can be embodied by one and the same item of hardware.

Claims

1. A power supply circuit, comprising:

a rectifier unit (D) for rectifying an AC mains voltage (Vm);
a first switch (T) coupled across first and second outputs of the rectifier unit (D); and
a control circuit (T1, T1', Rb) for controlling the first switch (T) in such a manner, that the first switch (T) is switched on when a voltage over the first switch (T) is substantially zero, the control circuit (T1, T1', Rb) including:
a second switch (T1, T1') having a main current path between a control terminal of the first switch (T) and the first output of the rectifier unit; and
a resistor (Rb) between a control terminal of the second switch (T1, T1') and the second output of the rectifier unit (D).

2. A power supply circuit as claimed in claim 1, wherein the first switch (T) is a thyristor.

3. A power supply circuit as claimed in claim 1, wherein the second switch (T1, T1') is a bipolar transistor.

4. A power supply circuit as claimed in claim 1, wherein a diode (D1) is connected between the control terminal of the first switch (T) and the second switch (T1, T1').

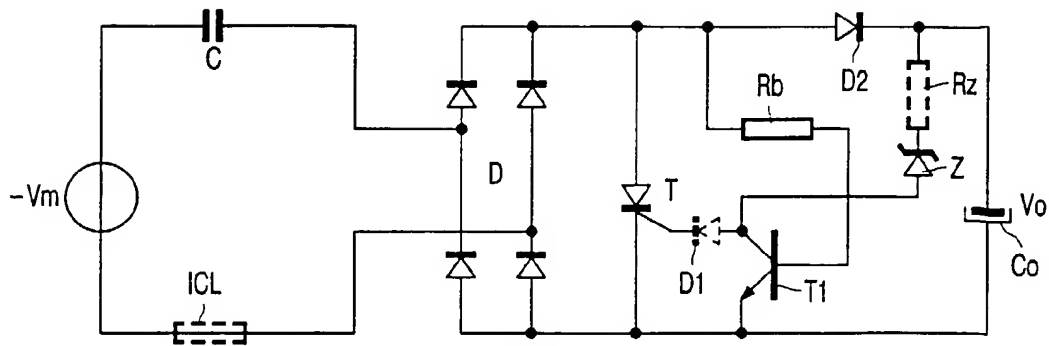


FIG. 1

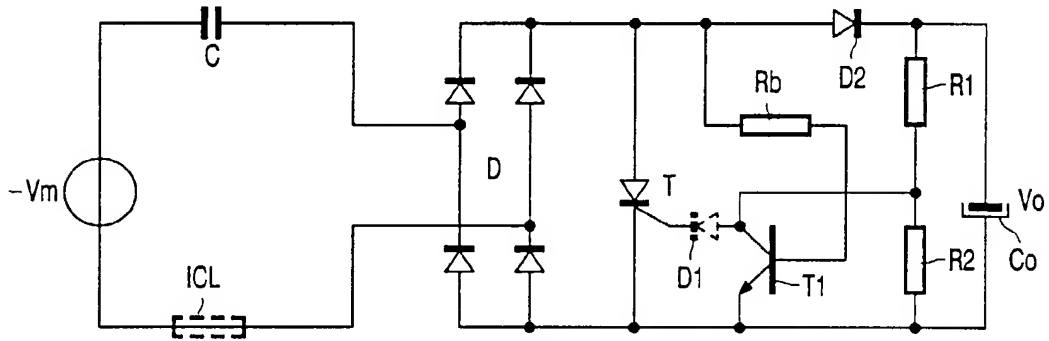


FIG. 2

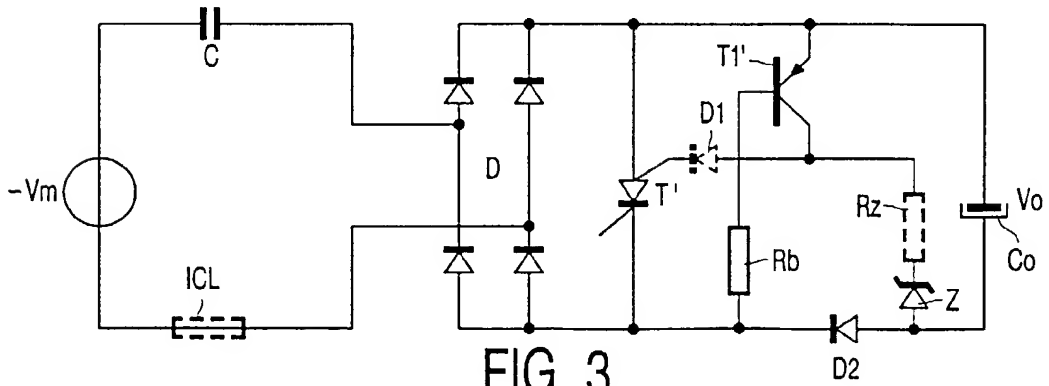


FIG. 3

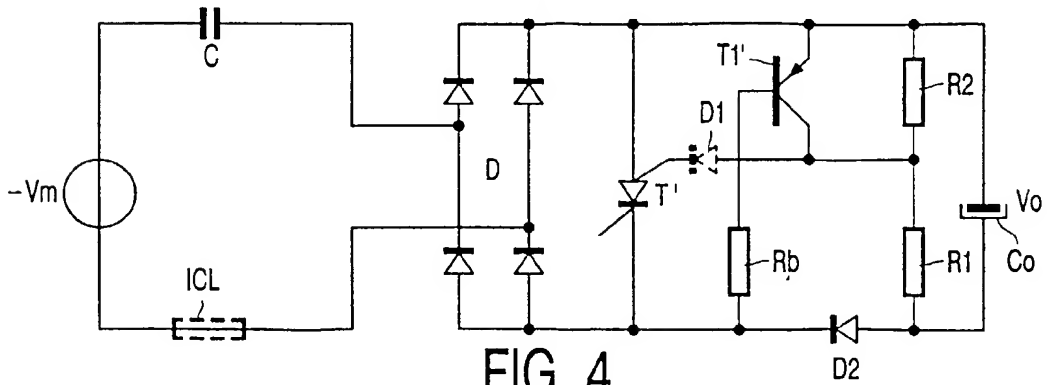


FIG. 4



European Patent
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EUROPEAN SEARCH REPORT

Application Number
EP 99 20 2816

DOCUMENTS CONSIDERED TO BE RELEVANT			
Category	Citation of document with indication, where appropriate, of relevant passages	Relevant to claim	CLASSIFICATION OF THE APPLICATION (Int.Cl.7)
X	DE 26 32 496 A (HAUBNER GEORG ING GRAD) 26 January 1978 (1978-01-26)	1-3	H02M3/137
A	* figure 2 *	4	
A	US 4 300 090 A (WEBER HAROLD J) 10 November 1981 (1981-11-10)	1-4	
A	US 5 430 637 A (BUCK DEAN C) 4 July 1995 (1995-07-04)	1-4	
D,A	EP 0 030 266 A (SICK OPTIK ELEKTRONIK ERWIN) 17 June 1981 (1981-06-17)	1-4	
D,A	EP 0 324 904 A (WEISS ALFONS AWE GMBH CO KG) 26 July 1989 (1989-07-26)	1-4	
The present search report has been drawn up for all claims			TECHNICAL FIELDS SEARCHED (Int.Cl.7)
			H02M
Place of search		Date of completion of the search	Examiner
THE HAGUE		2 May 2000	Lampe, S
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EPO FORM 1503 03 82 (P/4/001)

**ANNEX TO THE EUROPEAN SEARCH REPORT
ON EUROPEAN PATENT APPLICATION NO.**

EP 99 20 2816

This annex lists the patent family members relating to the patent documents cited in the above-mentioned European search report.
The members are as contained in the European Patent Office EDP file on
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02-05-2000

Patent document cited in search report		Publication date	Patent family member(s)		Publication date
DE 2632496	A	26-01-1978	NONE		
US 4300090	A	10-11-1981	NONE		
US 5430637	A	04-07-1995	JP	7284266 A	27-10-1995
EP 0030266	A	17-06-1981	DE	2949347 A	11-06-1981
EP 0324904	A	26-07-1989	DE	3801399 A	03-08-1989
			AT	84384 T	15-01-1993
			DE	3877342 A	18-02-1993

EPO FORM P0459

For more details about this annex : see Official Journal of the European Patent Office, No. 12/82